## **Frequency-Domain Sampling**

Recall the Fourier transform of an aperiodic discrete-time signal:

$$X(\omega) = \sum_{n = -\infty}^{\infty} x(n) e^{-j\omega n}$$

Evaluate this at equidistant samples  $\omega = \frac{2\pi}{N}k$ :

$$X(\frac{2\pi}{N}k) = \sum_{n = -\infty}^{\infty} x(n)e^{-j\frac{2\pi}{N}kn}, \qquad k = 0, 1, 2, ..., N-1$$

The summation can be subdivided:

$$X(\frac{2\pi}{N}k) = \dots + \sum_{\substack{n = -N \\ 2N-1 \\ + \sum_{\substack{n = N \\ n = N}} x(n)e^{-j2\pi kn/N} + \sum_{\substack{n = 0 \\ n = N}} x(n)e^{-j2\pi kn/N} + \dots$$
$$= \sum_{\substack{l = -\infty \\ l = -\infty}}^{\infty} \sum_{\substack{n = lN \\ n = lN}}^{\infty} x(n)e^{-j2\pi kn/N} \qquad k = 0, 1, 2, \dots, N-1$$

If we interchange the summations:

$$X(\frac{2\pi}{N}k) = \sum_{n=0}^{N-1} \left[\sum_{l=-\infty}^{\infty} x(n-lN)\right] e^{-j2\pi kn/N}$$

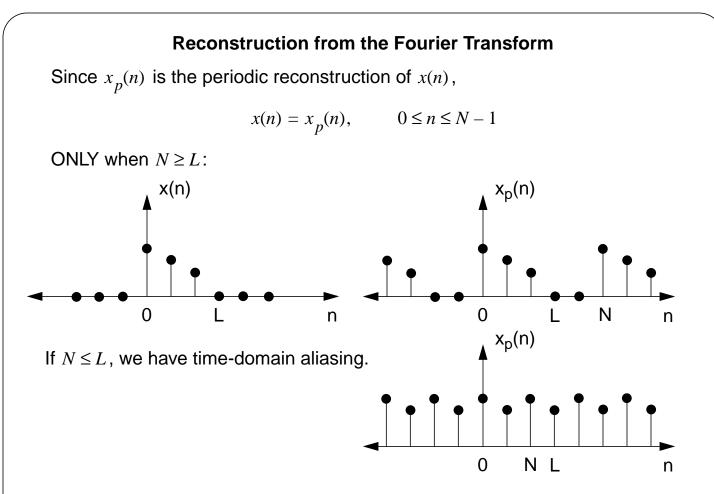
The signal

$$x_p(n) = \sum_{l = -\infty}^{\infty} x(n - lN)$$

is obtained by the periodic repetition of x(n) every N samples. Its Fourier Series is given by:

$$x_p(n) = \sum_{n=0}^{N-1} c_k e^{j2\pi kn/N}$$
 and  $c_k = \frac{1}{N} \sum_{n=0}^{N-1} x_p(n) e^{-j2\pi kn/N}$ 

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Assuming  $N \ge L$ ,

$$x(n) = \sum_{k=0}^{N-1} X(\frac{2\pi}{N}k)e^{j2\pi kn/N}$$

We can write the Fourier transform in terms of  $X(\frac{2\pi}{N}k)$ :

$$X(\omega) = \sum_{n=0}^{N-1} \left[ \frac{1}{N} \sum_{k=0}^{N-1} X(\frac{2\pi}{N}k) e^{j2\pi kn/N} \right] e^{-j\omega n}$$

This can be simplified to:

$$X(\omega) = \sum_{k=0}^{N-1} X(\frac{2\pi}{N}k)P(\omega - \frac{2\pi}{N}k), \qquad N \ge L$$

where

$$P(\omega) = \frac{\sin(\omega N/2)}{N\sin(\omega/2)}e^{-j\omega(N-1)/2}$$

What is the significance of this equation?

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## Frequency Domain Interpolation Via Zero-Padding

Suppose x(n) = 0 for n < 0 and  $n \ge L$ . Let us define  $x_p(n)$ :

$$x_p(n) = \begin{cases} x(n), & 0 \le n \le L - 1\\ 0, & L \le n \le N - 1 \end{cases}$$

We define the discrete Fourier transform (DFT) of x(n) as:

$$X(k) = \sum_{n=0}^{N-1} x(n) e^{-j2\pi kn/N}, \qquad k = 0, 1, 2, ..., N-1$$

and the inverse discrete Fourier transform (IDFT):

$$x(n) = \sum_{n=0}^{N-1} X(k) e^{j2\pi kn/N}, \qquad n = 0, 1, 2, ..., N-1$$

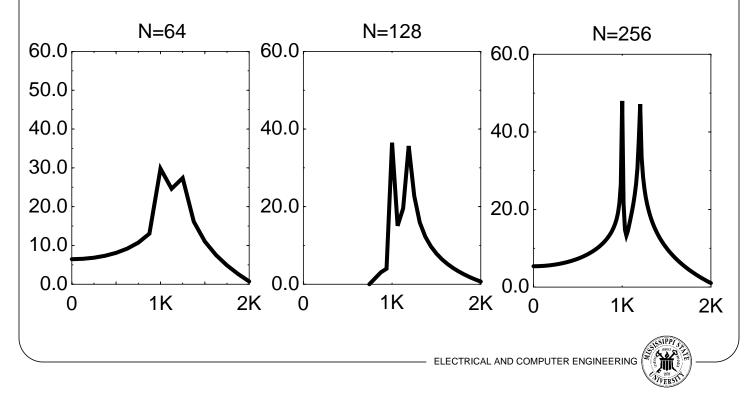
Clearly, if L < N, x(n) = 0 for  $L \le n \le N - 1$ .

Important questions:

What is the meaning of life?

Why  $N \gg L$ ?

Example:  $f_1 = 999.0 \text{ Hz}$ ,  $f_2 = 1199.0 \text{ Hz}$ ,  $f_s = 8000.0 \text{ Hz}$ 



## The DFT as a Linear Transformation

We can rewrite the DFT as follows:

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn} \qquad k = 0, 1, ..., N-1$$

$$x(n) = \sum_{n=0}^{N-1} X(k) W_N^{-kn} \qquad n = 0, 1, ..., N-1$$

where  $W_N = e^{-j2\pi/N}$ , which is an Nth root of unity.

We can express the DFT as a matrix operation by defining:

$$\boldsymbol{x}_{N} = \begin{bmatrix} x(0) \\ x(1) \\ \dots \\ x(N-1) \end{bmatrix} \qquad \boldsymbol{X}_{N} = \begin{bmatrix} X(0) \\ X(1) \\ \dots \\ X(N-1) \end{bmatrix}$$
$$\boldsymbol{W}_{N} = \begin{bmatrix} 1 & 1 & 1 & \dots & 1 \\ 1 & W_{N}^{1} & W_{N}^{2} & \dots & W_{N}^{N-1} \\ 1 & W_{N}^{2} & W_{N}^{4} & \dots & W_{N}^{2(N-1)} \\ 1 & W_{N}^{N-1} & W_{N}^{2(N-1)} & \dots & \dots \\ 1 & W_{N}^{N-1} & W_{N}^{2(N-1)} & \dots & W_{N}^{(N-1)(N-1)} \end{bmatrix}$$

and rewriting the DFT as:

$$\boldsymbol{X}_N = \boldsymbol{W}_N \boldsymbol{x}_N$$

and the inverse DFT as:

$$\boldsymbol{x}_N = \boldsymbol{W}_N^{-1} \boldsymbol{X}_N$$

## **DFT Matrix Operations**

Note that, given our definition of  $W_N$ , the IDFT can be expressed as:

$$\boldsymbol{x}_N = \frac{1}{N} \boldsymbol{W}^*{}_N \boldsymbol{X}_N$$

Therefore, we can equate IDFT equations and write:

$$\boldsymbol{W}_N^{-1} = \frac{1}{N} \boldsymbol{W}^*{}_N$$

which, in turn, implies that

$$\boldsymbol{W}_{N}\boldsymbol{W}^{*}{}_{N} = N\boldsymbol{I}_{N}$$

where  $I_N$  is an NxN identity matrix.

Therefore, the matrix  $W_N$  is an orthogonal (unitary) matrix, and its inverse exists.

More important questions:

If a tree falls in the forest...

Is  $W_N$  redundant?

Can we simply the number of computations required for  $X_N$ ?

